

# **Low frequency noise measurements in direct detection radiometers**

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***Receiver Gain Stability***

**5th Engineering Forum Workshop (Cagliari / Italy, 12 - 13 May 2011)**

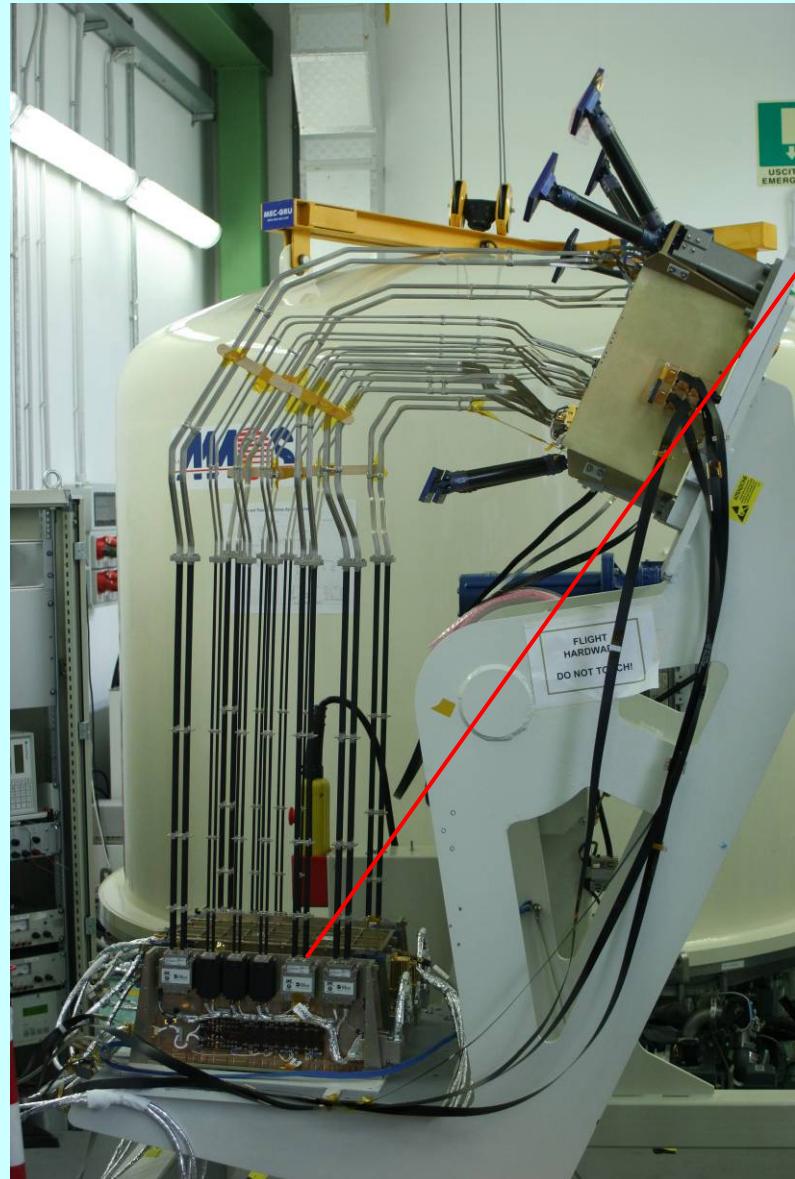


**RadioNet-FP7, 5th Engineering Forum Workshop (Cagliari, 12-13 May 2011)**



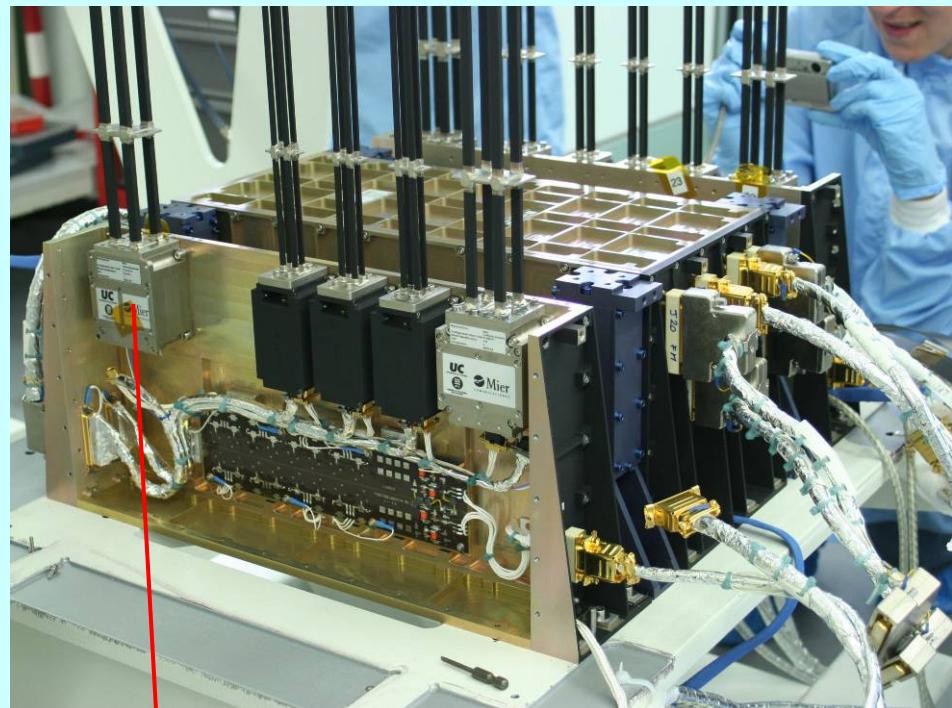
# Why we test 1/f noise?

- Planck mission (ESA): Low Frequency Instrument (LFI)
- To map spatial anisotropy in the Cosmic Microwave Background (CMB)
- Data with low 1/f noise to achieve scientific objectives
- Pseudo-correlation radiometers (to cancel 1/f noise)



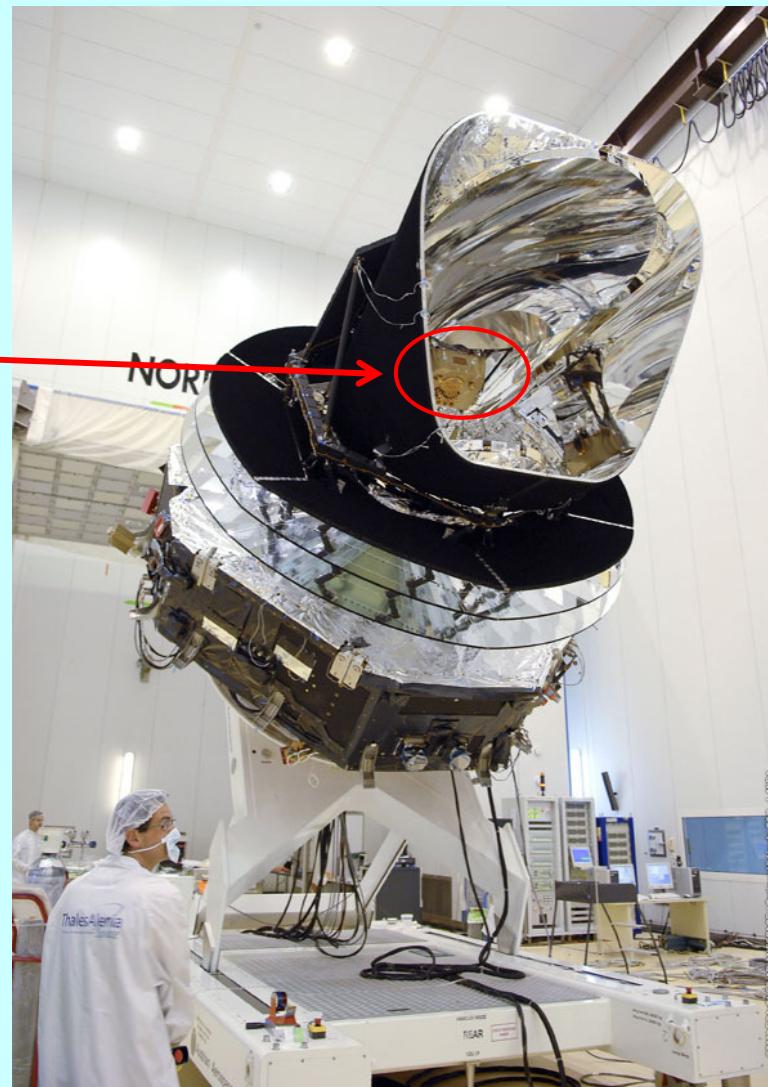
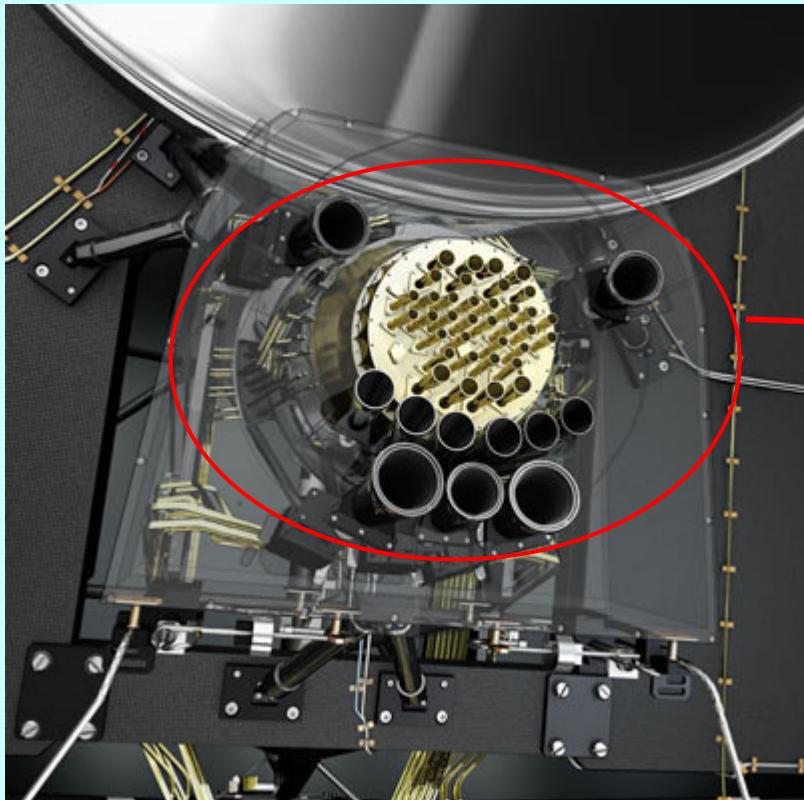
## Planck-LFI integration

Three Back End Modules at 44 GHz

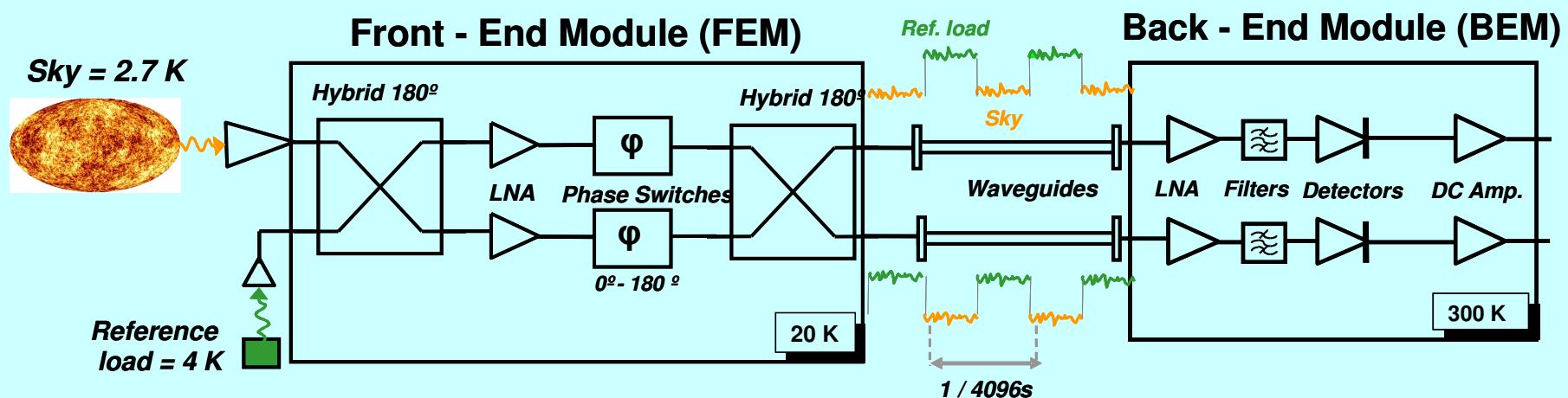


Two Back End Modules at 30 GHz

# Planck receivers (LFI + HFI)



# Planck-LFI radiometer scheme



# 1/f noise limitations in Planck-LFI

- Planck satellite rotation (scan):  $1 \text{ rpm} \cong 0.0166 \text{ Hz} = f_{\text{spin}}$
- Post-detection knee frequency  $f_k < f_{\text{spin}}$  (should be)
- For  $f_k > f_{\text{spin}} \Rightarrow$  mitigate 1/f effects by “destriping and map making algorithms”
- Small residual knee frequency (of  $\sim 0.1 \text{ Hz}$ )

# 1/f noise limitations in Planck-LFI

Strategies to mitigate 1/f noise:

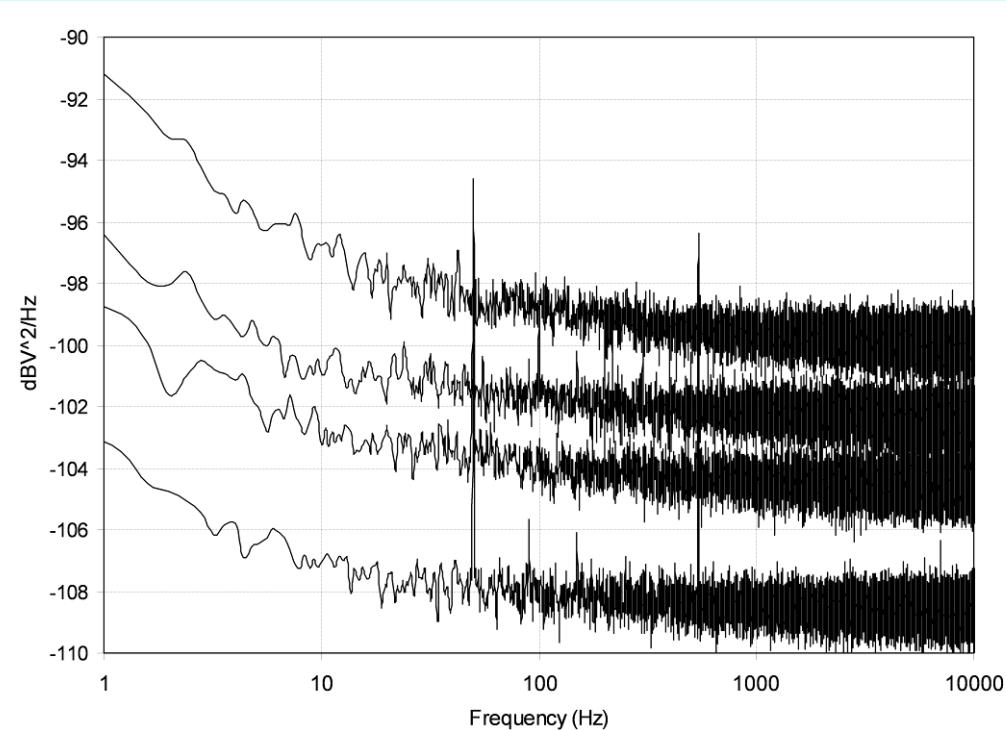
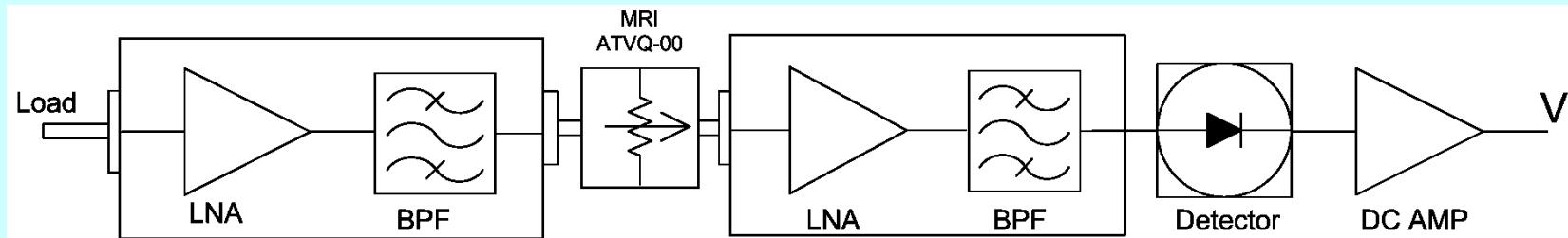
- Gain modulation (r factor applied in software) compensates different temperatures of the reference load (4 K) and sky ( 2.7 K):

$$r = \frac{T_{sys} + T_{sky}}{T_{sys} + T_{ref}} \implies V_{out} = A \left( T_{sky} + T_{sys} - r (T_{ref} + T_{sys}) \right) \approx 0$$

- Fast phase switching ( $f_{sw} \sim 4$  kHz) reduces the impact of 1/f fluctuations of BEM amplifiers ( $f_{sw} \gg f_{kBEM}$ ):

$$f_{kBEM} \ll 4 \text{ kHz}$$

## Example: Planck 44 GHz BEM 1/f noise results



Output spectra for RF  
relative input powers to the  
detector:

- 0 dB (bottom)
- 5.4 dB
- 8.2 dB
- 15 dB (top)

$f_{kBEM}$  increases slightly

$f_{kBEM} \sim 80 \text{ Hz} \ll 4 \text{ kHz}$



## Noise levels and test equipment

- Thermal noise spectral density ( $T_0 = 290 \text{ K}$ ;  $B = 1 \text{ Hz}$ )

$$S_n = k T_0 B = 4 \times 10^{-21} \text{ (Watt / Hz)} = -174 \text{ (dBm / Hz)}$$

- Signal Analyzer (HP 89410A) noise floor:  
typical  $\sim -165 \text{ dBm/Hz at 1kHz}$
- Lock In Amplifier (SR 830) typical input noise:

$$6 \text{ nV}/\sqrt{\text{Hz}} \text{ at 1kHz}$$

Input impedance:  $10 \text{ M}\Omega \Rightarrow$  Noise floor  $\approx -204 \text{ dBm/Hz at 1 kHz}$

## Conversion between units

Noise voltage spectral density  $v_n$  ( $V/\sqrt{Hz}$ )

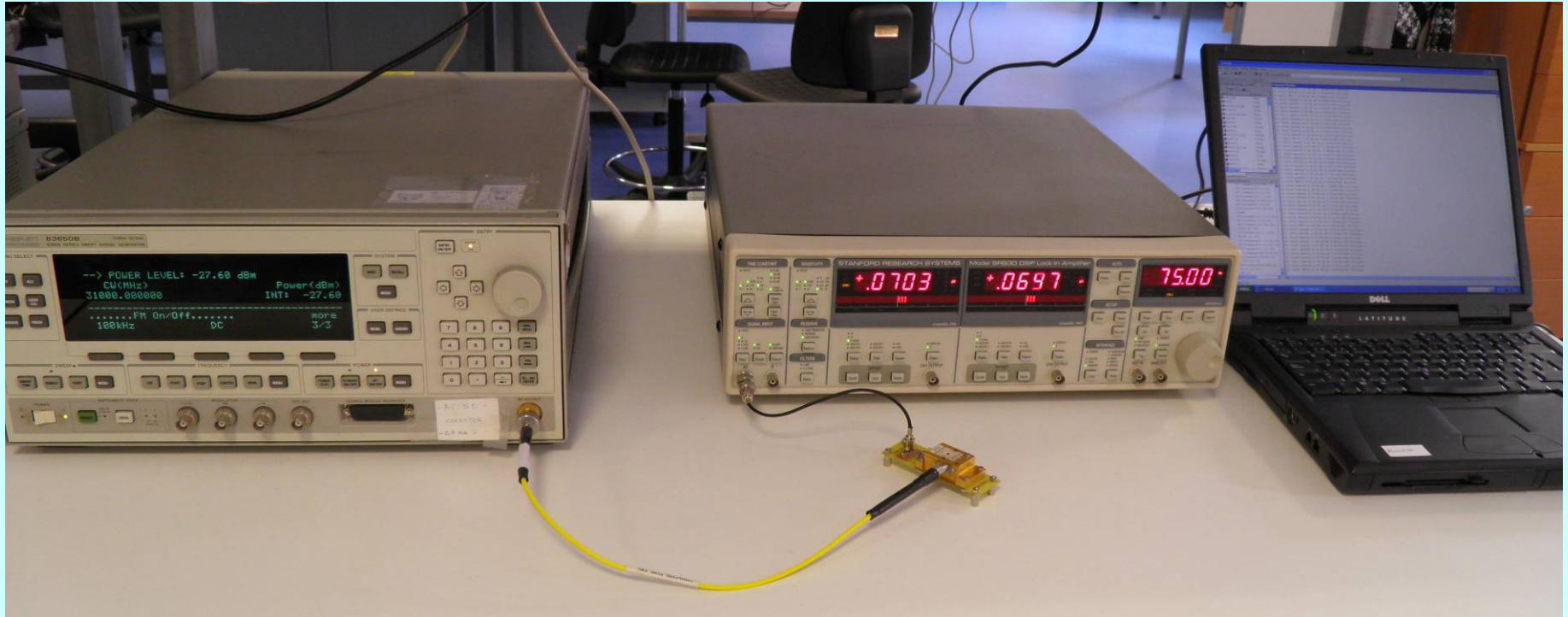
Noise power spectral density  $S_n$  ( $dBm/Hz$ )

$$S_n \text{ } (dBm/Hz) = 10 \log \left( \frac{v_n^2}{R_{in}} \right) + 30$$

Example:

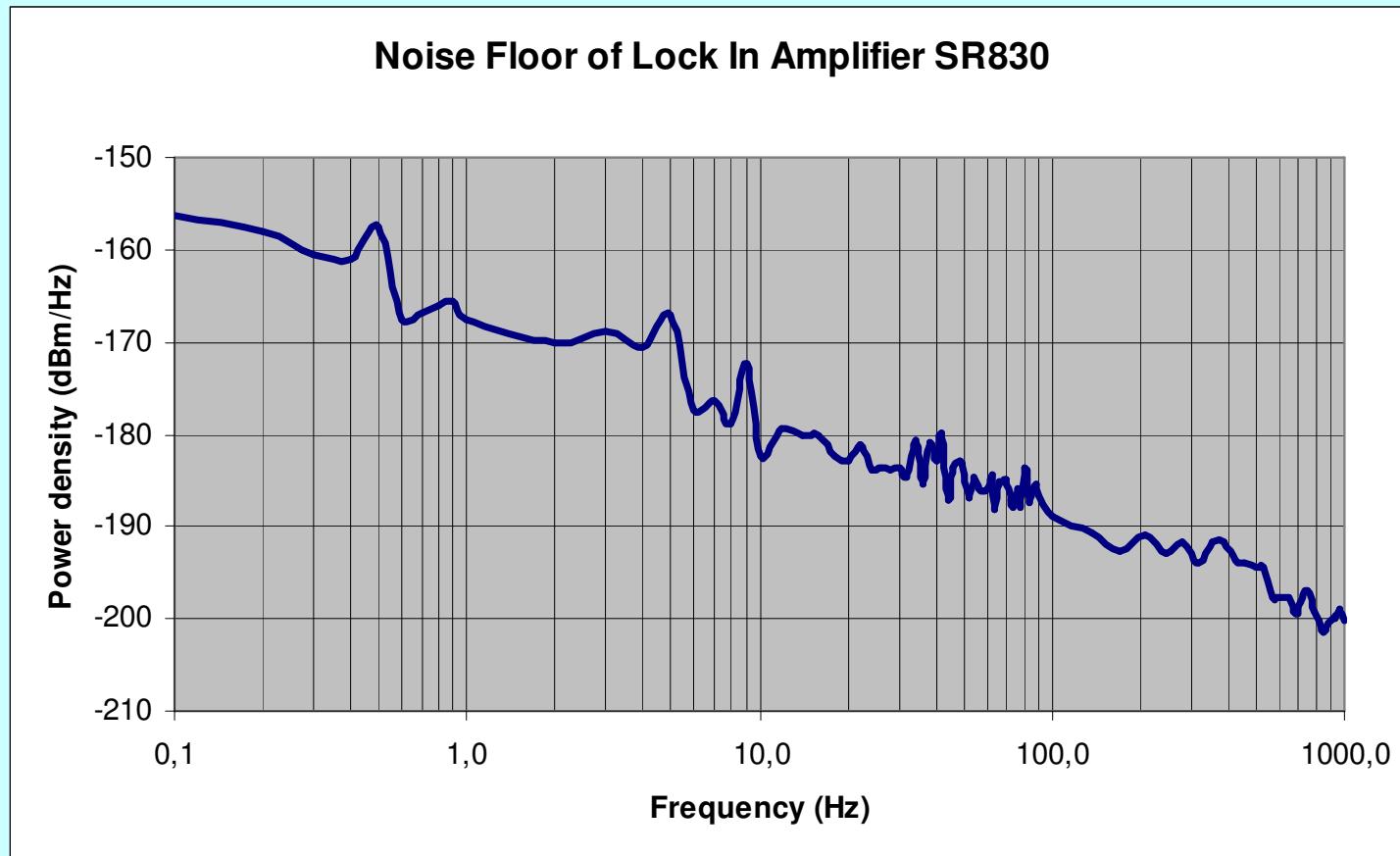
$$v_n = 6 \text{ } (nV/\sqrt{Hz}) \quad S_n = -204.4 \text{ } (dBm/Hz)$$

# Testing 1/f Noise with a Lock-In Amplifier SR830 (1 mHz - 102 kHz)



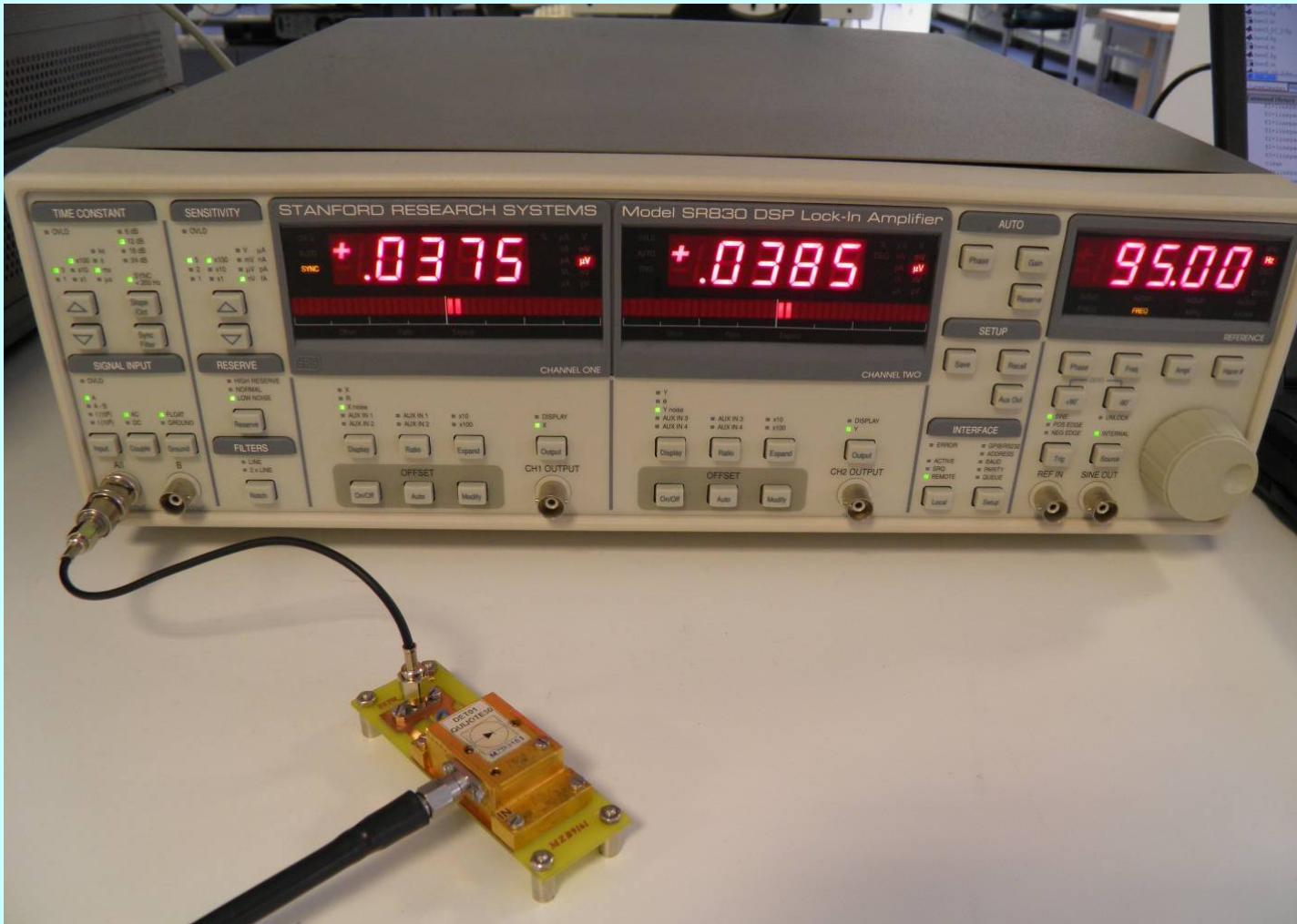
High sensitivity tests  $\Rightarrow$  Large time constant ( $\tau$ )  $\Rightarrow$  Large waiting and average times  $\Rightarrow$  typical  $\sim 9$  hours for 55 freq. (0.1 to 100 Hz): 550 samples

# Noise floor of Lock-In Amplifier system



Input load: Short ( $R_L = 0$ );  $\tau = 300$  ms; slope = 12 dB/oct; sensitivity = 1  $\mu$ V

# Testing noise with the Lock-In Amplifier



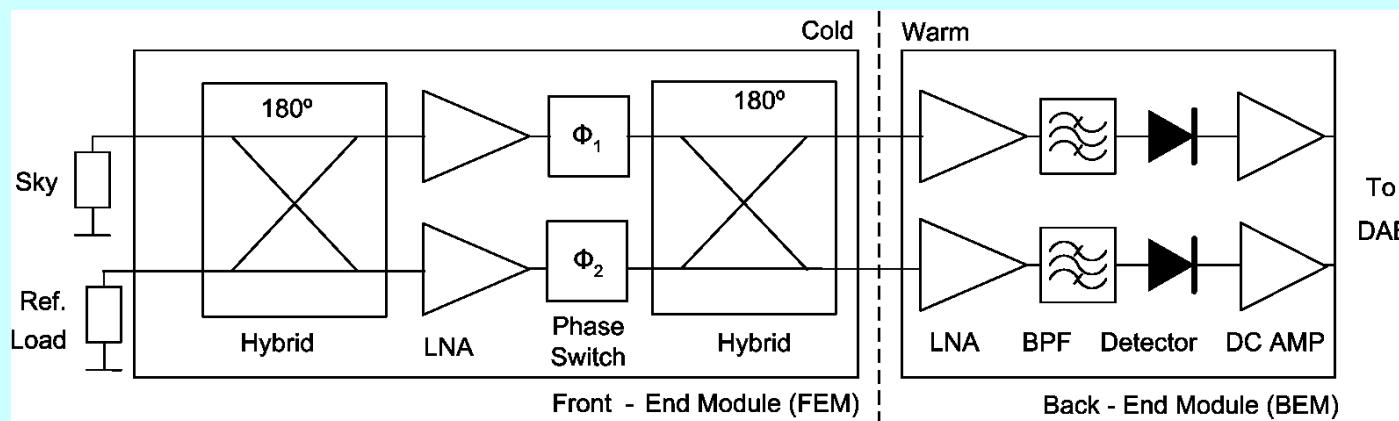
# Noise testing with Signal Analyzers



Signal Analyzer  
HP 89410A

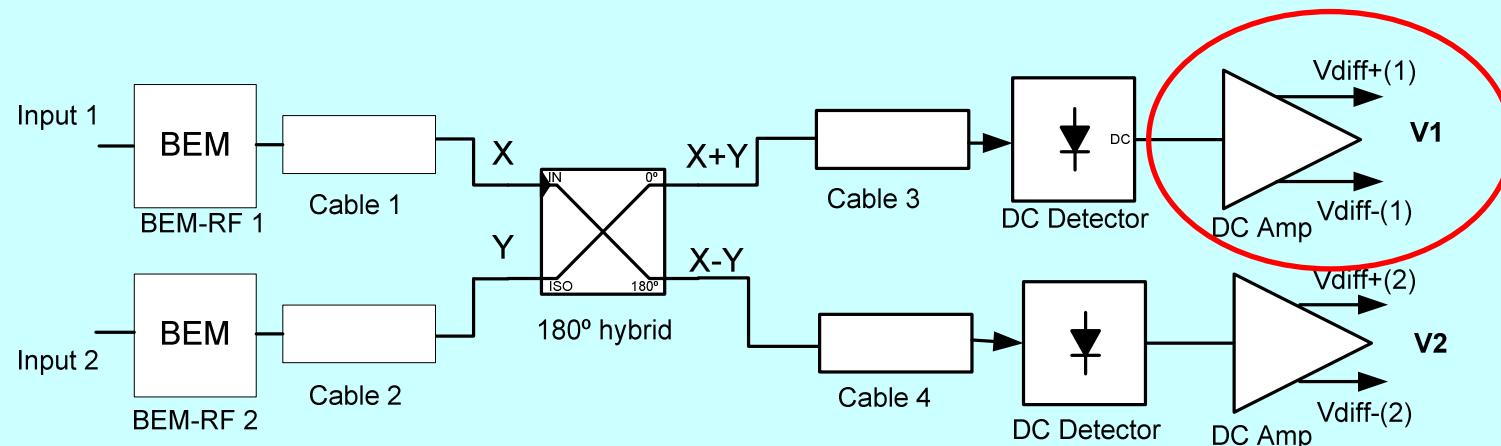
# 1/f noise contribution of radiometer subsystems

- DC amplifier
- Schottky diode detector (zero bias)
- LNA (Back End Module at RT)
- LNA cryogenic (Front End Module)



# DC amplifier (in BEM QUIJOTE-1 radiometer)

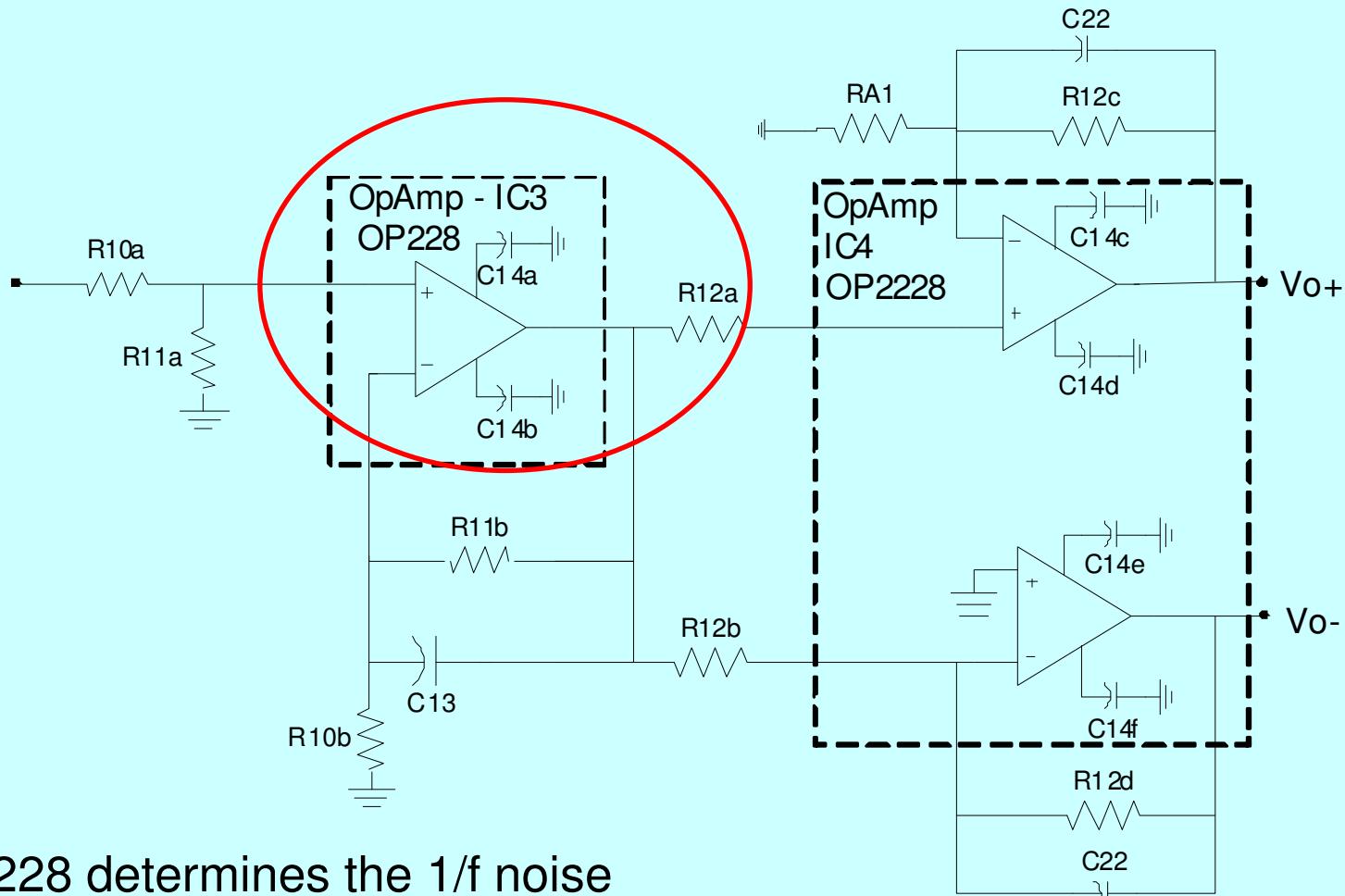
Bandwidth: 26 to 36 GHz



Output voltages are differential signals  
to meet EMC requirements and grounding integrity

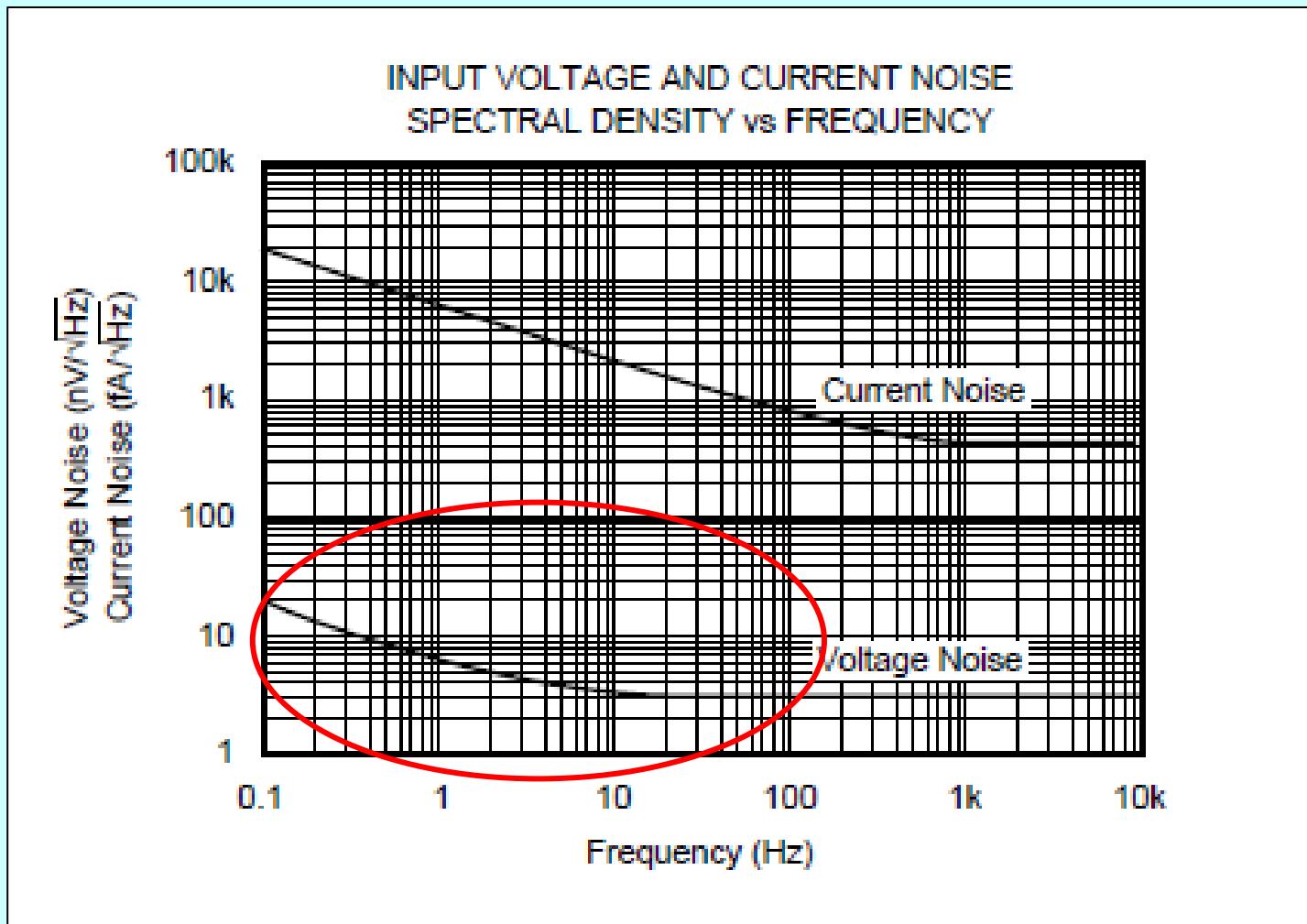
Voltage balanced gain = 580

# DC amplifier scheme



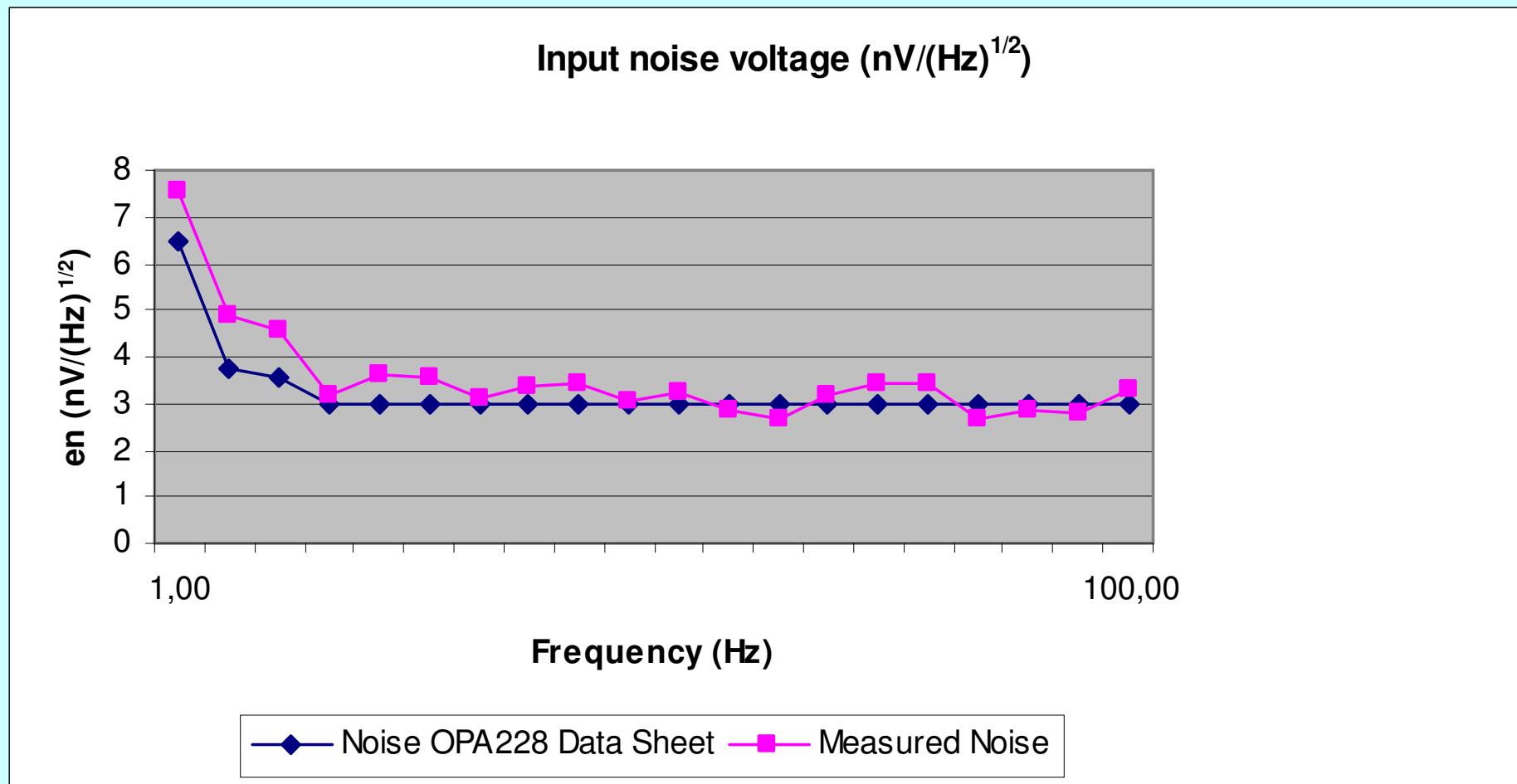
OPA228 determines the 1/f noise  
( $Z_{in} \approx 50.6 \text{ k}\Omega$ ;  $Z_{out} < 1 \Omega$ ;  $G_{DC} = 580$ )

# From OPA228 data sheet

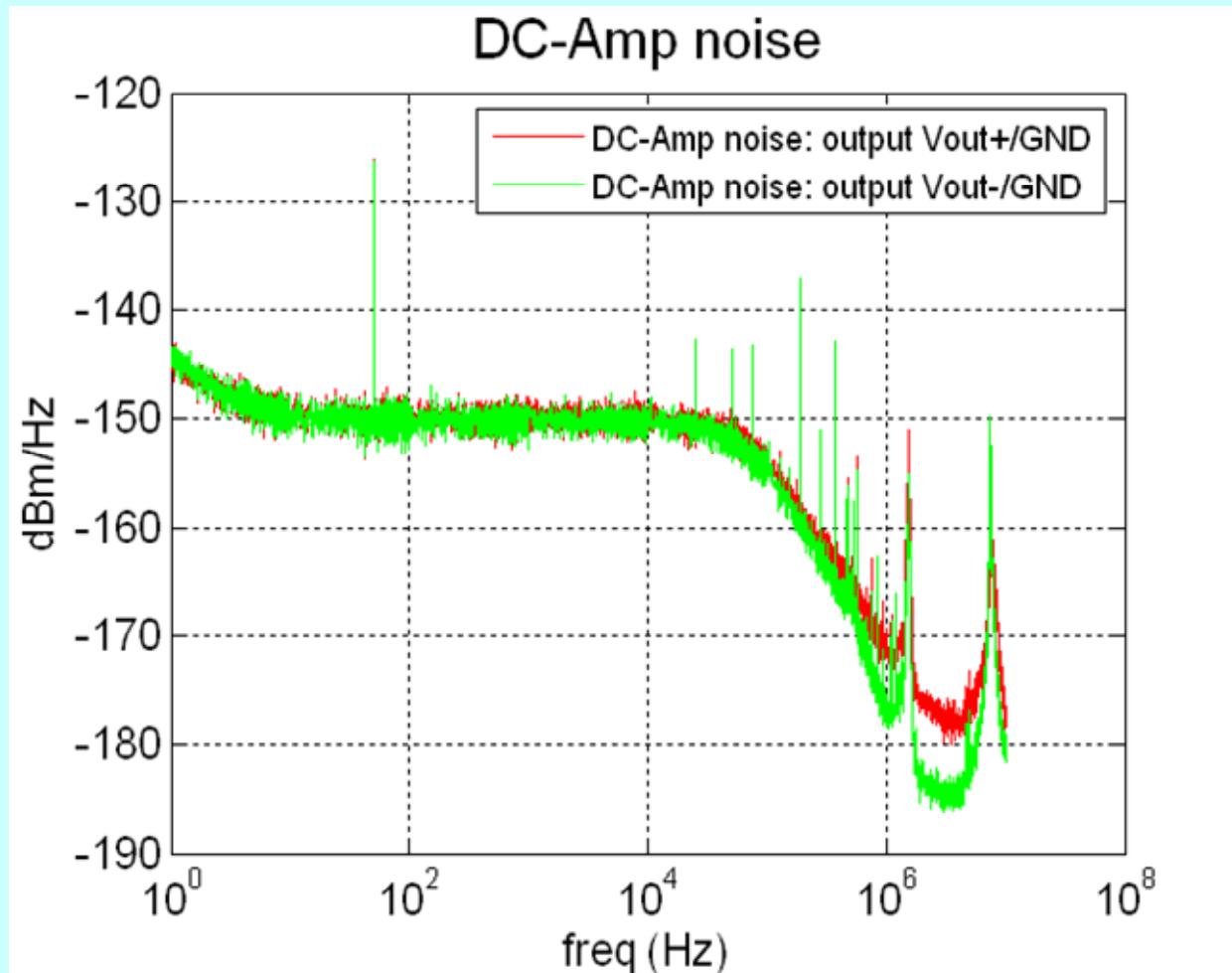


$3 \text{ } nV/\sqrt{\text{Hz}}$

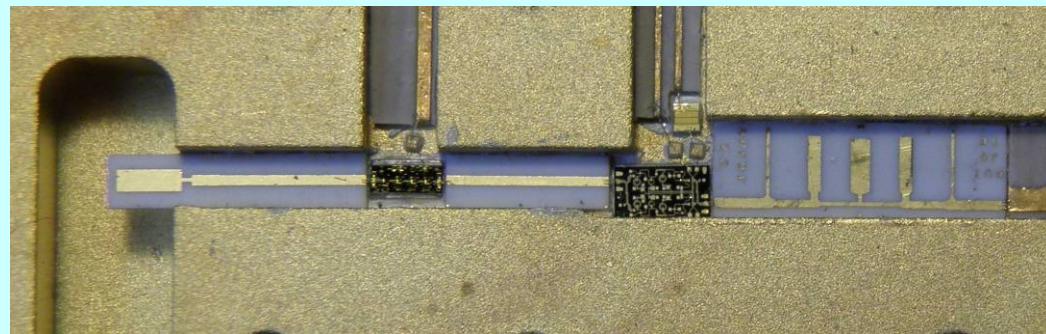
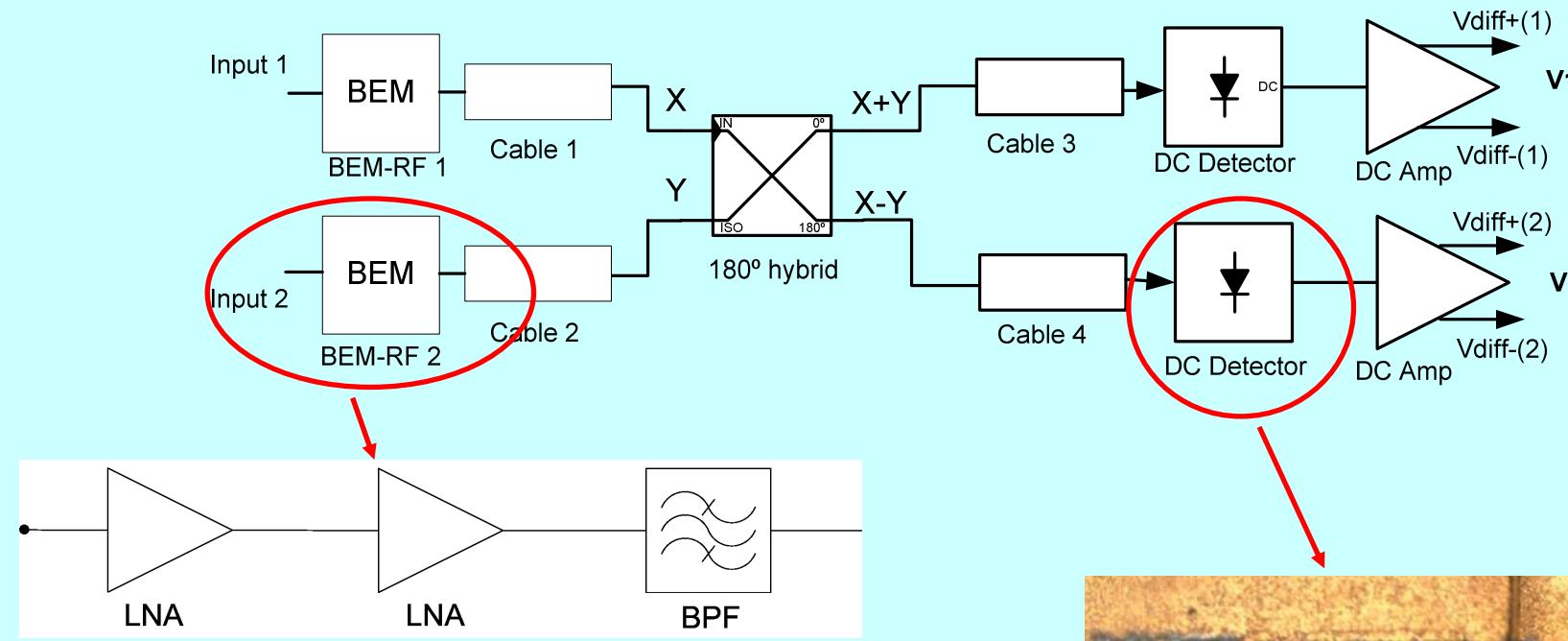
# Comparison OPA228 – Test with Lock-In Amp.



# DC amp Noise tests with Signal Analyzer HP-89410A

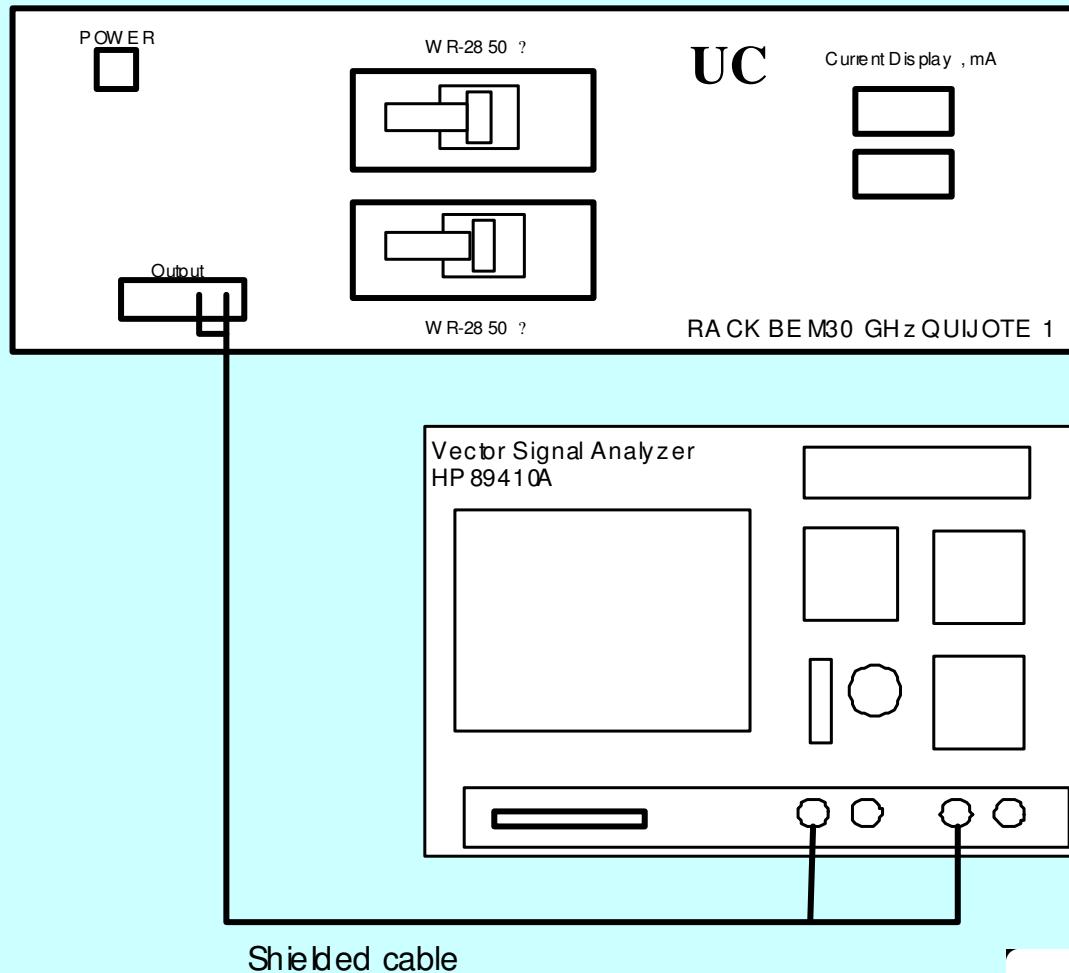


# Complete BEM 1/f noise tests (QUIJOTE-1: 26-36 GHz)

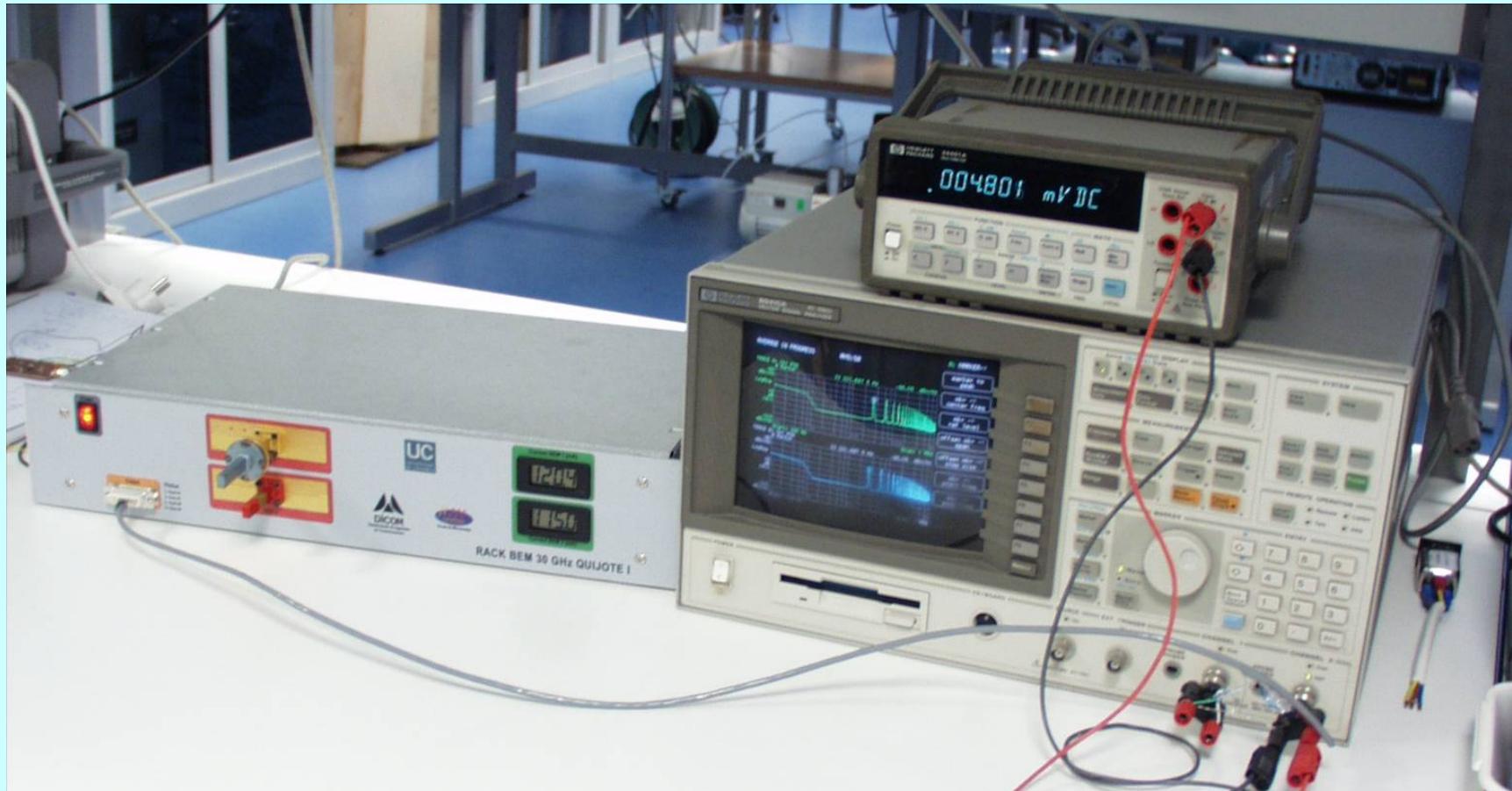


# Test set-up for BEM low frequency noise

**QUIJOTE-1  
26-36 GHz**



# Low frequency noise test with Signal Analyzer



# Low frequency noise in zero bias Schottky diode detector

Flicker and shot noise power spectral densities ( $\text{A}^2/\text{Hz}$ ):

$$S_{if} = k_f \frac{I_{DC}^a}{f^b} \quad S_{ishot} = 2q(I_{DC} + 2I_S)$$

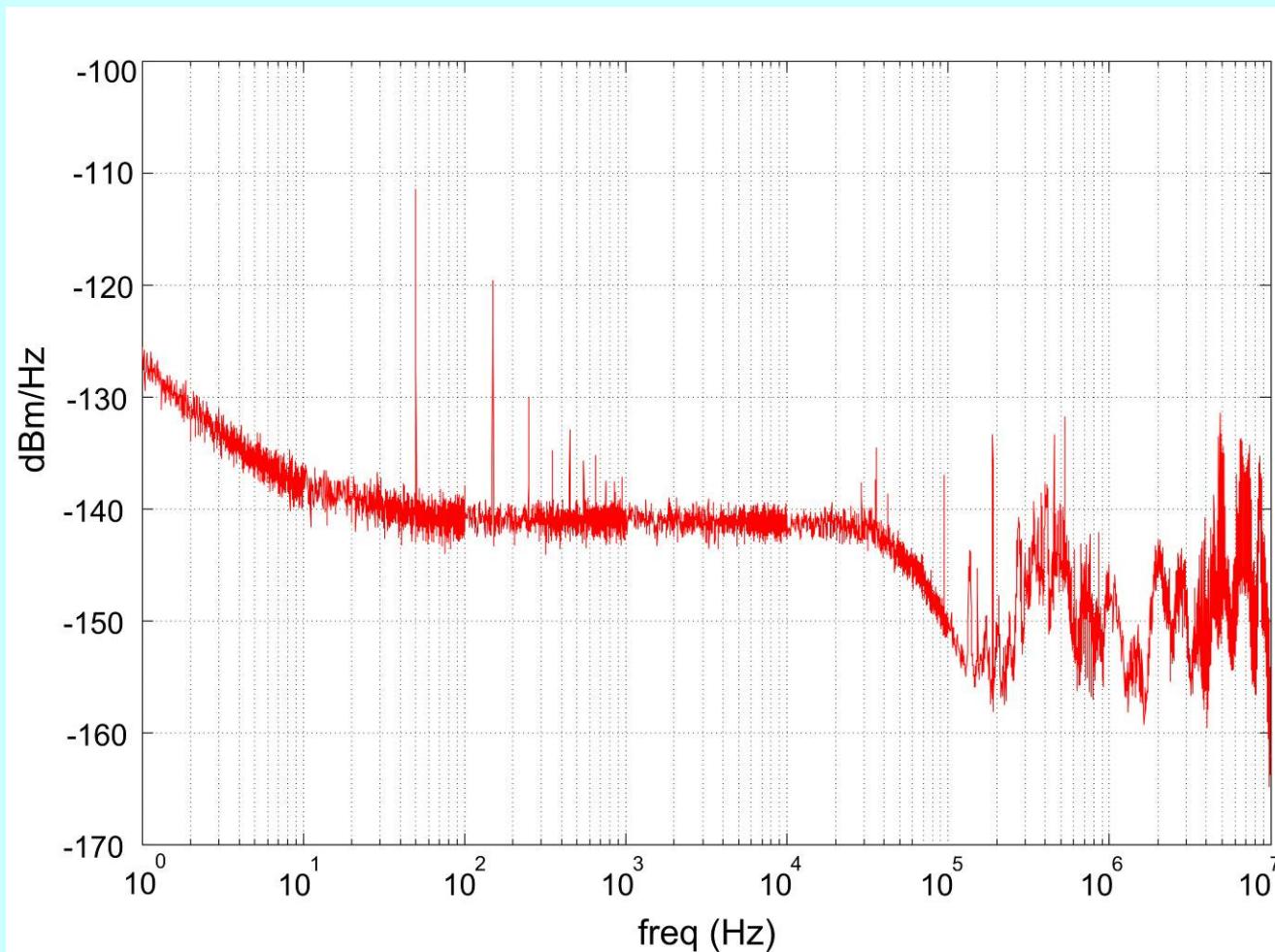
$I_{DC}$  = rectified DC forward current (proportional to RF power)

$I_S$  = diode saturation current

$q$  = electron charge

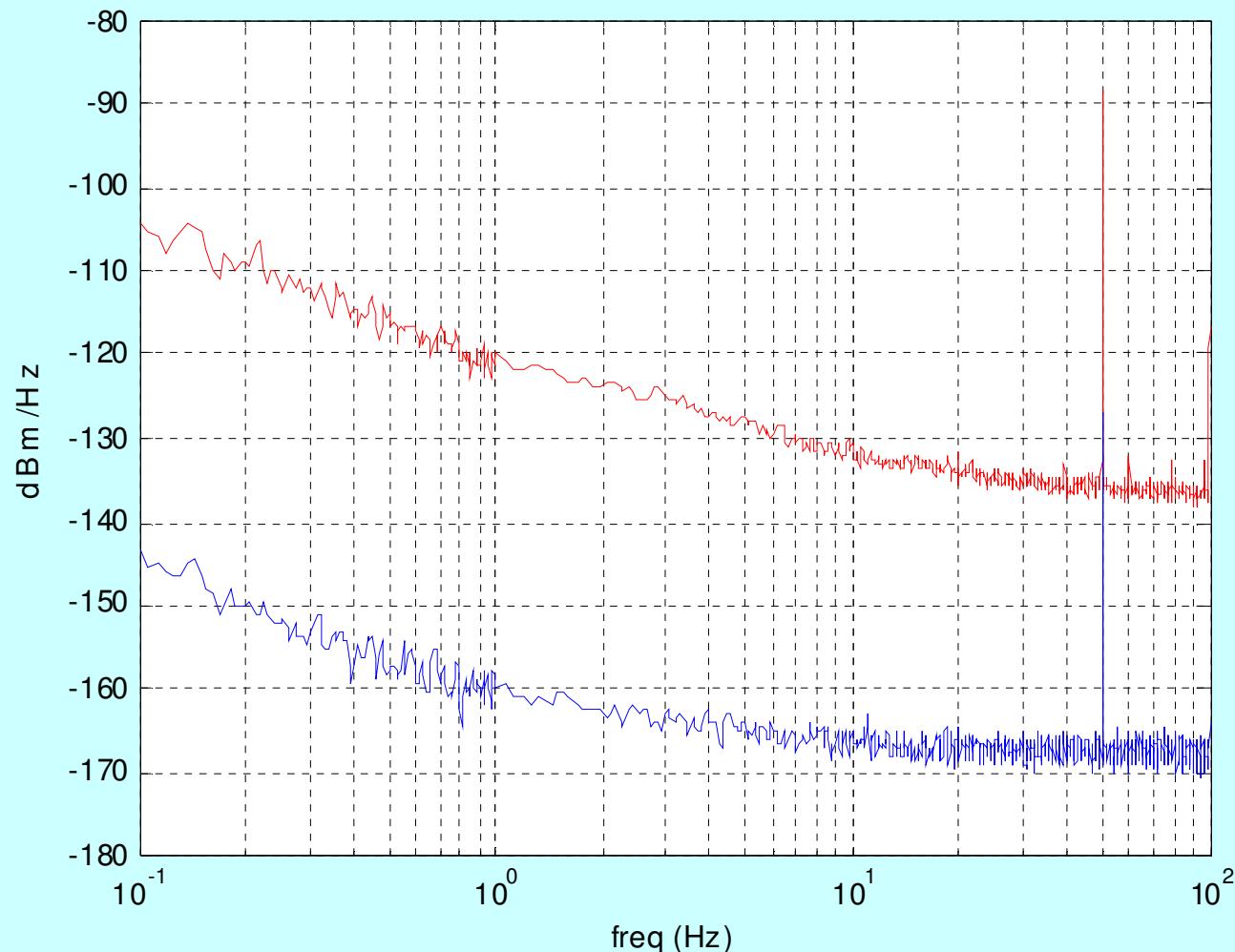
$k_f$ ,  $a$ ,  $b$  are fitting variables

# BEM (QUIJOTE-1) output noise



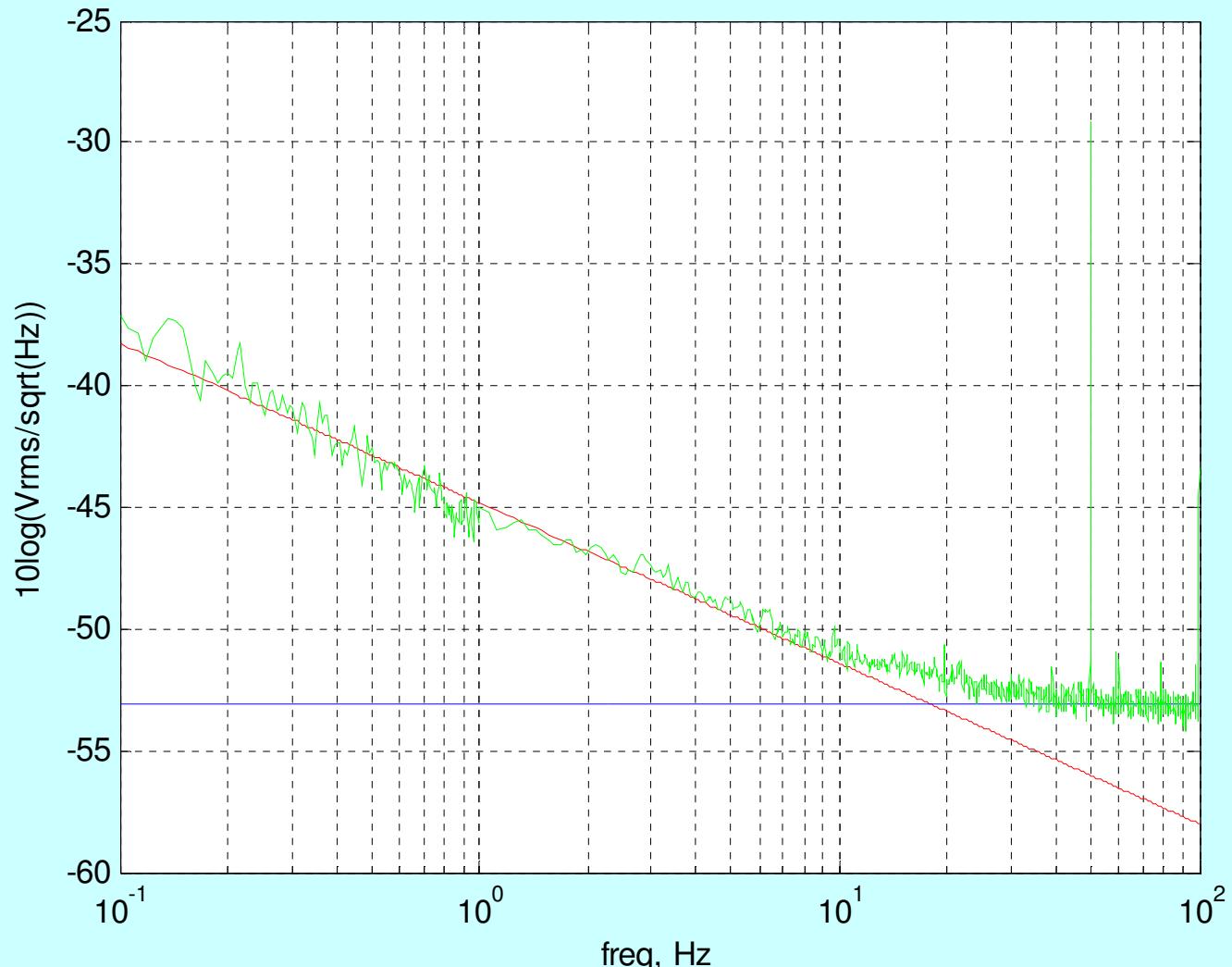
1/f knee frequency  $\sim 20$  Hz

# BEM (QUIJOTE-1) output noise and noise floor



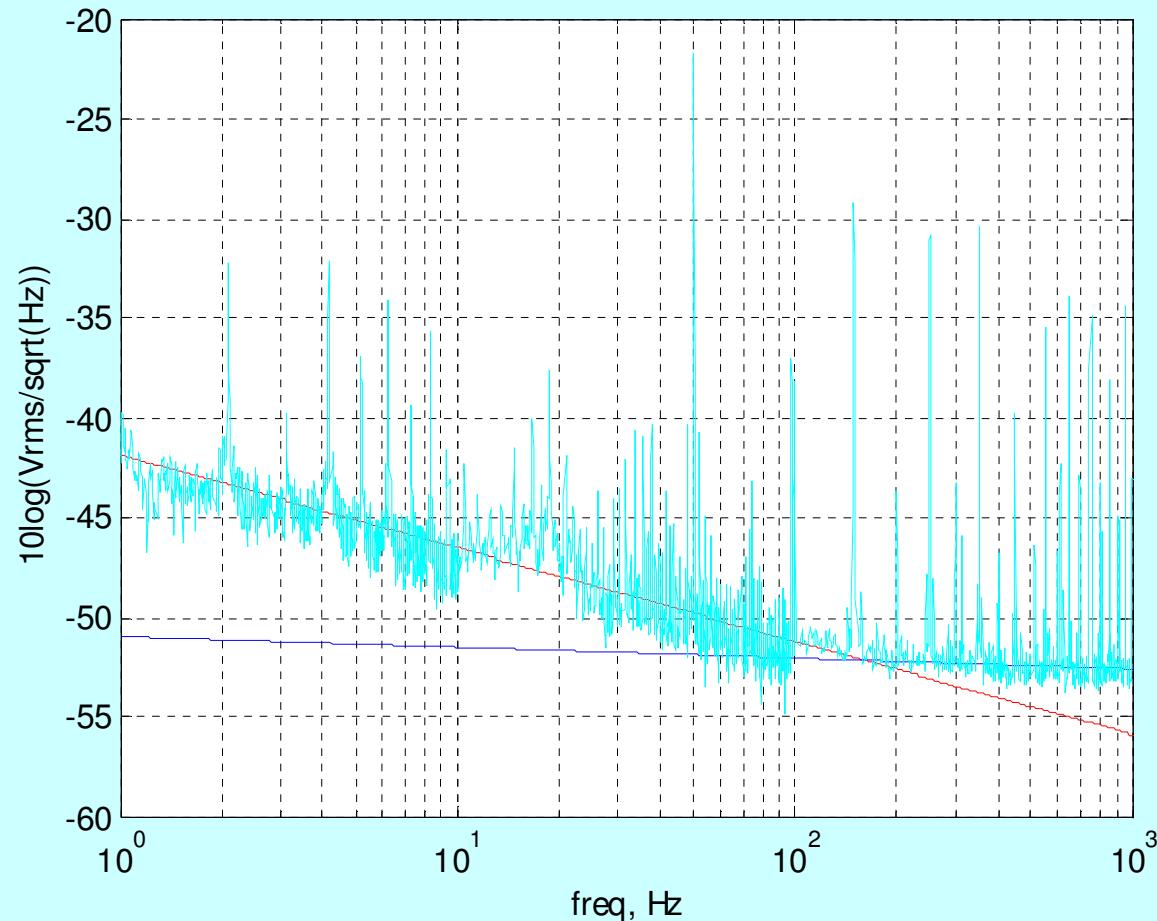
Test equipment: Signal Analyzer HP 89410A

# Knee frequency estimation by straight lines



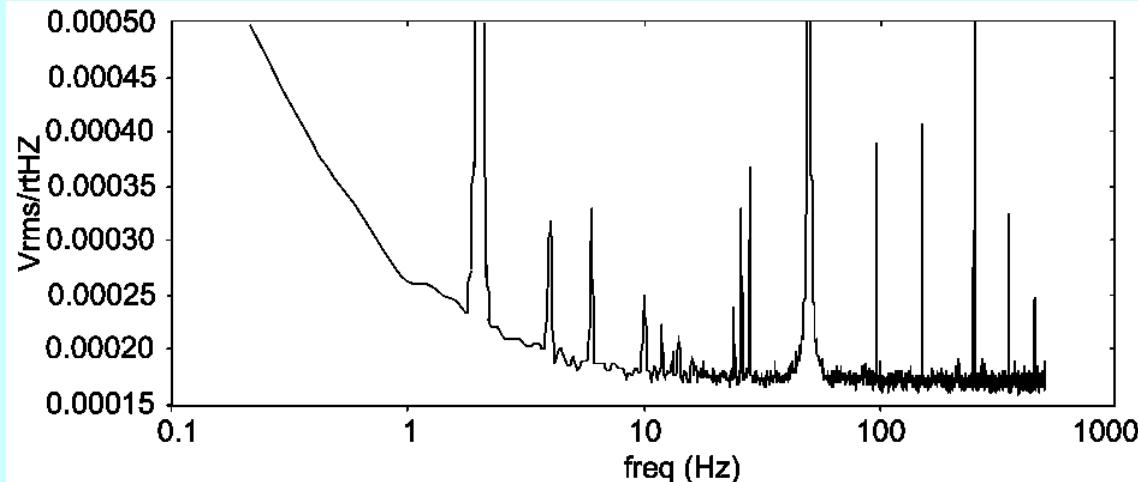
1/f knee frequency  $\sim 18$  Hz

# FEM-cryo + BEM (QUIJOTE-1) output noise

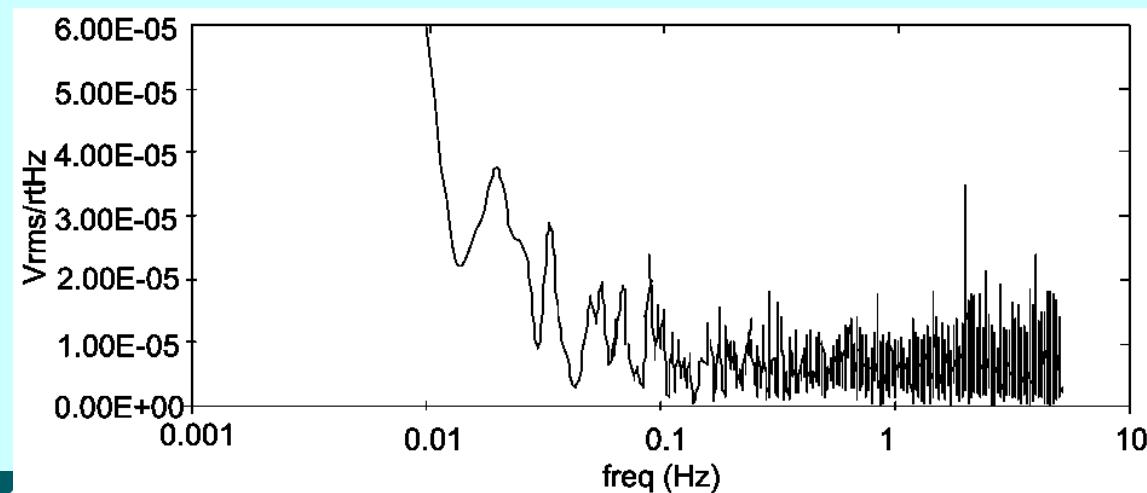


1/f knee frequency ~ 120 Hz

# Planck radiometer 30 GHz (prototype)



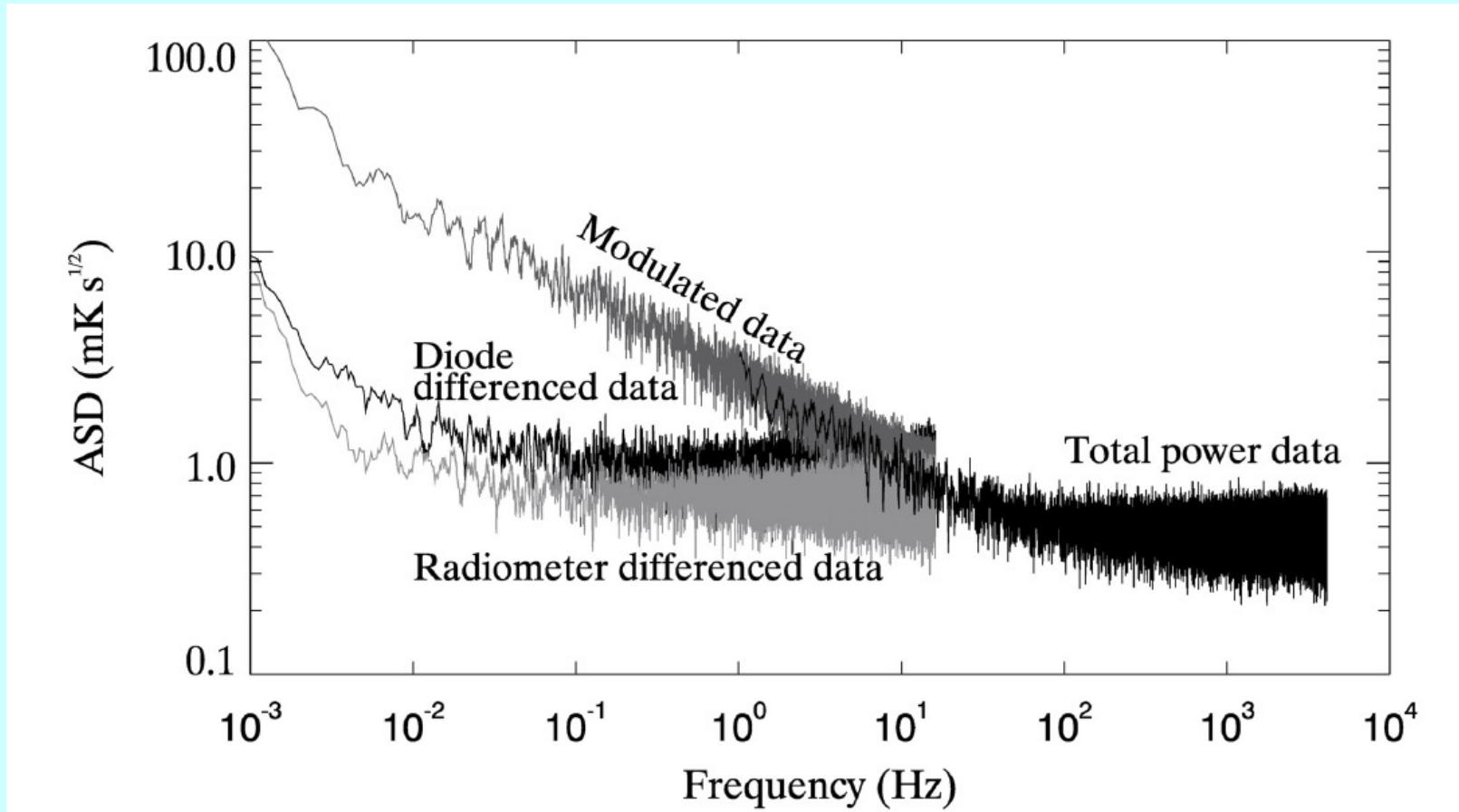
Unswitched



Switched

$f_{knee} \sim 50$  mHz

# 1/f noise reduction in Planck radiometers



Amplitude Spectral Densities of unswitched and differenced data streams.  
Reduction by 3 orders of magnitude of the 1/f knee frequency.  
(A. Menella et al., 2010, A&A, 520, A5)

# Conclusions

- 1/f noise degrades the quality of measured data.
- Cryogenic HEMT amplifiers (gain and noise temperature fluctuations) are the major source of 1/f noise.
- Pseudo-correlation differential radiometers can reduce the 1/f knee frequency by 3 orders of magnitude.

## References

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